

2.16

→ we calculated an overall noise factor of

$$F = 5.62 \text{ for this receiver}$$

total noise power referred to the input is

$$P_{ni} = kT_o B F_{ts}$$

$$= 1.38 \times 10^{-23} \cdot 290 \cdot 30 \times 10^3 \cdot 5.62$$

$$= 0.675 \times 10^{-15} \text{ W}$$

$$= -122 \text{ dBm}$$

∴ min required i/p signal is $-122 + 17 = -105 \text{ dBm}$

2.17 moving the LNA to before the lossy tx lines
results in

$$F = 1.42 \text{ (instead of 5.62)}$$

$$\text{now } P_{ni} |_{\text{dBm}} = -128 \text{ dBm} \quad (\text{instead of } -122 \text{ dBm})$$

min. detectable signal lowered by 6 dB
 clearly
 moving LNA
 $\therefore \text{min req. signal level} = -128 + 17 = -111 \text{ dBm}$ is better

(2)

2.18 A

1.) sig. power @ i/p of rx amplifier

$$P_s/\text{dB} = \text{EIRP}/\text{dB} + G_r/\text{dB} - L_{\text{sys}}/\text{dB} - L_{\text{path}}/\text{dB}$$

$$= 50 \text{ dBm} + 3 - 3 - 145 = -95 \text{ dBm}$$

2.) noise power spectrum referenced to i/p of rx amp

$$S_n = \frac{kT}{2} F_R = 12.6 \times 10^{-21} \frac{\text{W}}{\text{Hz}}$$

6.31

3.) SNR @ RX amp. o/p

→ noise power at RX o/p is

$$P_n = S_n \times 2 \times \text{BW}$$

$$= 12.6 \times 10^{-21} \cdot 2 \cdot 100 \times 10^3$$

$$= 2.53 \times 10^{-15} \text{ W} = -116 \text{ dBm} = P_n/\text{dB}$$

$$\therefore \text{SNR}_{\text{dB}} = P_s/\text{dB} - P_n/\text{dB} = -95 - (-116)$$

$$= 21 \text{ dB}$$

3

3.1

A) Just plug into $P_r \approx P_t G_t G_r h_t^2 h_r^2 / L_{sys} d^4$

$$\begin{aligned} P_r / \text{dB} &= P_t / \text{dB} + G_t / \text{dB} + G_r / \text{dB} + 20 \log(h_t) + 20 \log(h_r) \\ &\quad - L_{sys} / \text{dB} - 4 \times 10 \log(d) \\ &= -87.5 \text{ dBm} \end{aligned}$$

B) \Rightarrow just plug into the standard (ideal) range eqn.

$$P_r = \frac{P_t G_t G_r \lambda^2}{(4\pi d)^2} = -71 \text{ dBm} \quad (\lambda = 0.25 \text{ m})$$

C)

$$(3.16) \Rightarrow 2 A_{dir} \left| \sin \left(\frac{2\pi h_t h_r}{\lambda d} \right) \right| \leftarrow \text{subst given vals}$$

0.0754

0.0753

approx. v. good in this case

(4)

3.5

A) Using Hata model

$$\alpha(h_{ue}) = 3.2 \left(\log(11.75 h_{ue}) \right)^2 - 4.97 \text{ dB}$$

$$= 0.225 \text{ dB} \quad (\text{for } h_{ue} = 1.6 \text{ m})$$

$$\underbrace{\angle_{50}(\text{urban})}_{\substack{\text{median path} \\ \text{loss}}} / \text{dB} = 69.55 + 26.16 \log(1200) - 13.82 \cdot \log(30)$$

$$+ (44.9 - 6.55 \log(30)) \log(12) - 0.225$$

$$= 167 \text{ dB}$$

$$\text{B) } P_r / \text{dB} = 10 \cdot \log \left(\frac{15}{0.001} \right) \text{ dBm} + 7 + 2 - 4 - 167$$

$$= -121 \text{ dBm}$$

$$\text{C) } P_r / \text{dB} = 10 \log \left(\frac{15}{0.001} \right) + 7 + 2 - 4 - 40 \log(12 \times 10^3)$$

$$= +20 \log(30) + 20 \log(1.6)$$

$$= -82.8 \text{ dBm}$$

$\xrightarrow{\text{Eg. 3.25}}$ predicts a RX'd power $\sim 10,000X$ greater
 (v. optimistic)

(5)

3.8 repeat 3.5, but using Lee model
convert to imperial units

$$h_t = 98.4 \text{ ft}, h_r = 5.25 \text{ ft}, d = 7.46 \text{ mi}$$

Eg. (3.50) gives

$$\alpha_c = 20 \log\left(\frac{98.4}{100}\right) + 20 \log\left(\frac{5.25}{10}\right) - 30 \log\left(\frac{1260}{850}\right)$$

$$+ \left[10 \log\left(\frac{15}{1 \text{ mW}}\right) - 40 \right] + (7 - 8.15) + (2 - 2.15)$$

$$= -9.77 \text{ dB}$$

from Table 3.1 on pg. 93

$$P_{1-\text{mile}} = -77 \text{ dBm} \quad \nu = 4.8$$

$$\therefore P_{r,50} = -77 - 10 \times 4.8 \log\left(\frac{7.46}{1}\right) - 9.77 - 4$$

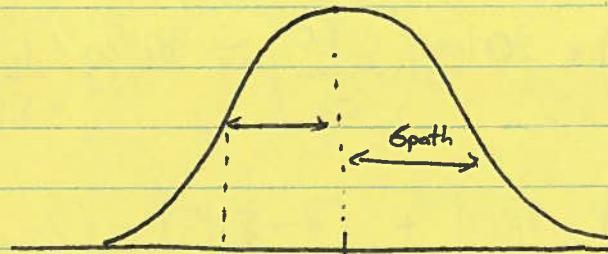
$$= -133 \text{ dBm}$$

(6)

3.10)

$f_m|_{dB}$: fade margin, probability that received signal will the amount below average median that can still be processed by my receiver

$$f_m|_{dB} = \overline{P_r}|_{dB} - p_{rl|dB}$$



$$p_{rl|dB} = (\overline{P_r}|_{dB} - \overline{P_r}|_{dB} - f_m|_{dB})$$

P_Q : probability that power available to rx will exceed the min. set by the fade margin

$$P_Q = \Pr [P_r|_{dB} > p_{rl|dB}] = Q\left(\frac{p_{rl|dB} - \overline{P_r}|_{dB}}{\sigma_{path}}\right)$$

$$= Q\left(\frac{(\overline{P_r}|_{dB} - f_m|_{dB}) - \overline{P_r}|_{dB}}{\sigma_{path}}\right) = Q\left(-\frac{f_m|_{dB}}{\sigma_{path}}\right)$$

(7)

3.13

$$P_{ni} = kTB_F = 3.19 \times 10^{-15} W = -115 \text{ dBm} = P_n |_{\text{dB}}$$

R

noise power referred to input

for SNR of 15 dB need min. received power

$$\text{of } P_r |_{\text{dB}} = -115 + 15 = -100 \text{ dBm}$$

$$\Pr[P_r |_{\text{dB}} > P_r |_{\text{dB}}] = 0.9 = Q\left(\frac{P_r |_{\text{dB}} - \bar{P}_r |_{\text{dB}}}{\sigma_{\text{path}}}\right)$$

$$= Q\left(-\frac{100 \text{ dBm} - \bar{P}_r |_{\text{dB}}}{8}\right) = 0.9$$

$$\Rightarrow \bar{P}_r |_{\text{dB}} = -89.7 \text{ dBm}$$

Lee model from 3.12

$$\alpha_c = 20 \log\left(\frac{100}{100}\right) + 20 \log\left(\frac{5}{10}\right) - 30 \log\left(\frac{1800}{850}\right) \\ + (44 - 40) + (4 - 8.15) + (12 - 2.15) = -16.1$$

$$P_{r,50} |_{\text{dB}} = -70 \text{ dBm} - 10 \times 3.68 \log\left(\frac{d}{1}\right) - 16.1 \text{ dB} - 2 = -89.7$$

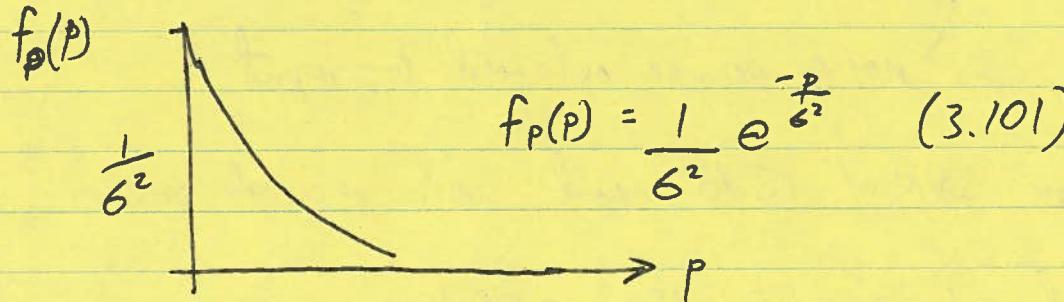
 $P_{\text{r-mile}} |_{\text{dB}}$

v

$$= \cancel{1.78 \text{ dBm}} \quad d = 1.11 \text{ mi} = 1.78 \text{ km}$$

⑧

3.16 \rightarrow probability density fn. of power in multipath environment is



$$\text{avg. } m_p = \sigma^2 = \bar{P}_r$$

$$\begin{aligned} \Pr[p_r > \bar{P}_r] &= \int_{\sigma^2}^{\infty} \frac{1}{\sigma^2} e^{-\frac{p}{\sigma^2}} dp \\ &= -e^{-\frac{p}{\sigma^2}} \Big|_{\sigma^2}^{\infty} = -0 - (-e^{-1}) \\ &= e^{-1} = 0.368 \end{aligned}$$

$$3.17 \quad \Pr[SNR > SNR_{min}] = 0.95 = \Pr\left[\frac{p_r}{P_n} > \frac{p_{r,min}}{P_n}\right] = \Pr[p_r > p_{r,min}]$$

$$\begin{aligned} &= \int_{p_{r,min}}^{\infty} \frac{1}{\sigma^2} e^{-\frac{p}{\sigma^2}} dp = e^{-\frac{p_{r,min}}{\sigma^2}} = 0.95 \\ &\sigma^2 = \bar{P}_r = 19.5 p_{r,min} \end{aligned}$$

$$\overline{SNR} = 19.5 SNR_{min}$$

$$\overline{SNR}|_{dB} = 12.9 dB + SNR_{min}|_{dB}$$

$$3.18 \quad \langle t_k \rangle = 4.86 \mu s$$

$$\langle t_k^2 \rangle = 25.4 \times 10^{-12} s^2$$

$$\sigma_d = \sqrt{\langle t_k^2 \rangle - \langle t_k \rangle^2} = 1.36 \mu s$$

$$B_{coh} = \frac{1}{56d} = 148 \text{ kHz}$$

\Rightarrow baseband $\Rightarrow 148 \text{ kbps}$

\Rightarrow but passband reqs. are 2x as great $\therefore \sim 74 \text{ kbps}$

$$3.19 \quad A) \quad 1.96 \text{ Hz} \Rightarrow \lambda = 0.158 \text{ m}$$

$$v = 60 \text{ mph} \Rightarrow 26.8 \text{ m/s}$$

$$f_d = \frac{v}{\lambda} = 170 \text{ Hz} : \text{max Doppler shift}$$

signals arrive at $-5^\circ * 85^\circ$ with respect to direction of travel

$$f_{d1} = f_d \cos(-5) = 169 \text{ Hz}$$

$$f_{d2} = f_d \cos(85) = 14.8$$

$T_{sig} \ll t_{u-n} : \text{fading is slow}$

$$f_{d1} - f_{d2} = 154 \text{ Hz}$$

B) \therefore period of fading cycle is $t_{u-n} = \frac{1}{154} = 6.48 \text{ ms}$

(10)

$$3.20 \quad \frac{G_{\text{path}}}{U} = \frac{9}{4.2} = 2.14$$

for area coverage of 90% need $P_{\text{sens}}(R) \approx 0.75$

(Table 3.21 pg. 137)

$$0.75 = \Pr [P_r(R)|_{\text{dB}} > P_{\text{sens}}|_{\text{dB}}] = Q\left(-\frac{f_m|_{\text{dB}}}{G_{\text{path}}}\right) = Q(-0.67)$$

$$\begin{aligned} f_m|_{\text{dB}} &= 0.67 \cdot G_{\text{path}} \\ &= 6.03 \text{ dB} \end{aligned}$$