

L13: Error Detection and Correction



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Outline

- Basic (channel) coding ideas
- Error detection (Backward error correction)
 - single parity
 - interleaved parity (2-D parity)
 - internet checksum
 - polynomial codes
- Effectiveness and Error Models
- (Forward) Error correction
 - cyclic codes, block codes, convolutional codes, iterative codes

Types of Coding

- Some options
 - Line Coding
 - spectrum control
 - timing
 - basic error detection
 - Channel Coding
 - error detection
 - error correction
 - error prevention (combined detection & decoding)



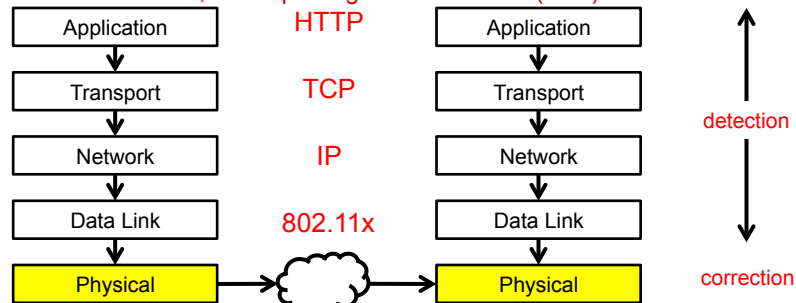
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Channel Coding

- Add in redundancy
- Two basic ideas, used in combination
 - error detection: recognizes an error in a frame (request re-send)
 - ARQ
 - error correction: finds error and corrects it (no need to re-send)
 - more desirable, but requires greater overhead (FEC)



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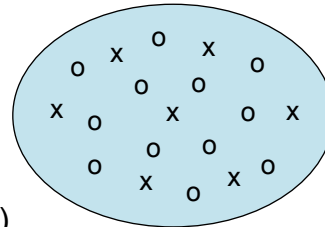
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Basic Ideas and Nomenclature

- Data consists of k bits
 - 2^k possible messages
- Add m redundant bits to this
- **Codeword** of $n = m + k$ bits
 - $2^{m+k} = 2^n > 2^k$ possible strings
 - But only 2^k are valid!
 - (n,k) codes e.g.: $(2,1)$
- Thus coded messages (*codewords*) are separated in signal space
 - Hamming distance, d : # of bit positions that differ
- **Code rate**: $r = k/n$
 - $\frac{1}{2}, \frac{3}{4}$

x = codeword
o = noncodewords



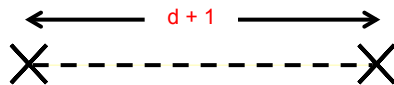
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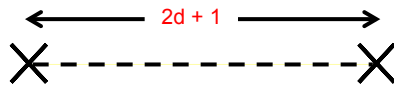
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Detection and Correction Basic Example

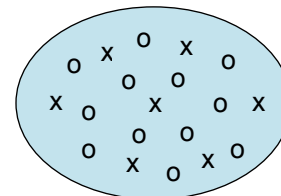
- To detect d errors: need Hamming distance of $d+1$



- To correct d errors: need Hamming distance of $2d+1$



- e.g.
 - 0000000000
 - 0000011111
 - 1111100000
 - 1111111111
- $d_{\min} = 5$
- detect up to 4 errors
- correct up to 2 errors
- only one at a time

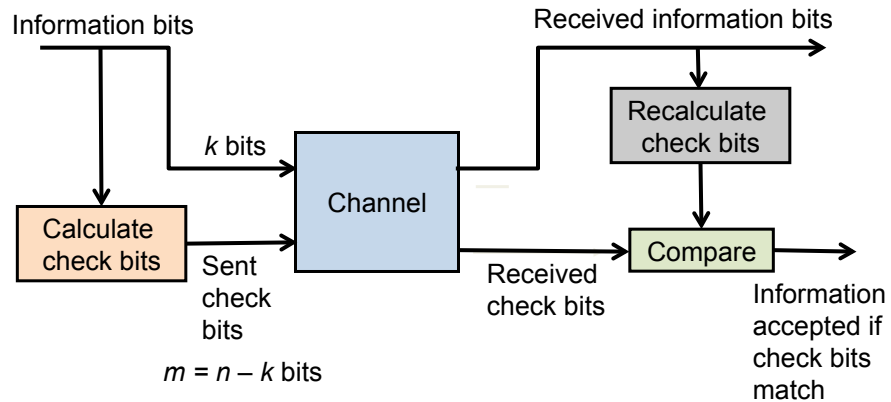


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System-Level Implementation



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Simple Detection: Single Parity Code

- Information (7 bits): (0, 1, 0, 1, 1, 0, 0)
- Parity Bit: $b_8 = 0 + 1 + 0 + 1 + 1 + 0 = 1$
- Codeword (8 bits): (0, 1, 0, 1, 1, 0, 0, 1)
- If single error in bit 3 : (0, 1, 1, 1, 1, 0, 0, 1)
 - # of 1's = 5, odd
 - Error detected
- If errors in bits 3 and 5: (0, 1, 1, 1, 0, 0, 0, 1)
 - # of 1's = 4, even
 - Error not detected

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Single Parity Check: Formally

- Append an overall **parity check** to k information bits
 - Info Bits: $b_1, b_2, b_3, \dots, b_k$
 - Check Bit: $b_{k+1} = b_1 + b_2 + b_3 + \dots + b_k \text{ modulo } 2$
 - Codeword: $(b_1, b_2, b_3, \dots, b_k, b_{k+1})$
- All codewords have even # of 1s
- **Redundancy**: Single parity check code adds 1 redundant bit per k information bits: overhead = $1/(k + 1)$
- **Coverage**
 - All error patterns that change an odd # of bits are detectable
 - All even-numbered patterns are undetectable
- Parity bit used in ASCII code

Effectiveness: Random Error Vector Model

- **Effectiveness**: Probability system fails to detect error
- Dependent on **error model**
 - Random Error Vector
 - Random Bit Error
 - Burst
- Random Error Vector
 - **n-bit vector**, e , represents error pattern
 - $e_i = 0$ -> no error in position i
 - $e_i = 1$ -> an error in position i
- 2^n possible combinations
 - Assumes all possibilities **equally likely**
 - 50% chance of even number of errors
 - Therefore...????

What If Bit Errors are Random?

- Many transmission channels **introduce bit errors at random**, independent of each other, with probability p
- **Some error patterns are more probable than others:**
 - For example, if $p = 0.1$

$$P[10000000] = p(1 - p)^7 = 0.0478$$

$$P[11000000] = p^2(1 - p)^6 = 0.0053$$

- In any worthwhile channel $p < 0.5$, and so $p/(1 - p) < 1$
- **It follows (can you show this?)** that patterns with 1 error are more likely than patterns with 2 errors and so forth
- What is the probability that an undetectable error pattern occurs?

Effectiveness: Random Bit Error Model

- Undetectable error pattern if even # of bit errors:

$$\begin{aligned} P[\text{error detection failure}] &= P[\text{undetectable error pattern}] \\ &= P[\text{error patterns with even number of 1s}] \end{aligned}$$

$$= \binom{n}{2} p^2 (1 - p)^{n-2} + \binom{n}{4} p^4 (1 - p)^{n-4} + \dots$$

- **Example:** Evaluate above for $n = 32$, $p = 10^{-3}$

$$\begin{aligned} P[\text{undetectable error}] &= \binom{32}{2} (10^{-3})^2 (1 - 10^{-3})^{30} + \binom{32}{4} (10^{-3})^4 (1 - 10^{-3})^{28} \\ &\approx 496 (10^{-6}) + 35960 (10^{-12}) \approx 4.96 (10^{-4}) \end{aligned}$$

- For this example, roughly 1 in 2000 error patterns is undetectable

Two-Dimensional Parity Check (Interleaving)

- More parity bits to improve coverage
- Arrange information as **rows**
- Add single parity bit to each **row**
- Add a final **“parity” row**
- Used in early error control systems

1	0	0	1	0		0
0	1	0	0	0		1
1	0	0	1	0		0
1	1	0	1	1		0
1	0	0	1	1		1

Last column consists of check bits for each row

Bottom row consists of check bit for each column

Error-Detecting Capability

1	0	0	1	0		0
0	0	0	0	0		1
1	0	0	1	0		0
1	1	0	1	1		0
1	0	0	1	1		1

One error

1	0	0	1	0		0
0	0	0	0	0		1
1	0	0	1	0		0
1	0	0	1	1		0
1	0	0	1	1		1

Two errors

1	0	0	1	0		0
0	0	0	1	0		1
1	0	0	1	0		0
1	0	0	1	1		0
1	0	0	1	1		1

Three errors

1	0	0	1	0		0
0	0	0	1	0		1
1	0	0	1	0		0
1	0	0	0	1		0
1	0	0	1	1		1

Four errors
(undetectable)

Arrows indicate failed check bits

- 1, 2, or 3 errors can always be detected
- Not all patterns >4 errors can be detected

Other Error Detection Codes

- Many applications **require very low error rate**
- Need codes that detect the vast majority of errors
- **Single parity** check codes do not detect enough errors
- **Two-dimensional** codes require too many check bits
- The following error detecting codes **used in practice**:
 - Internet Check Sums
 - CRC Polynomial Codes

Internet Checksum

- Several Internet protocols (e.g. IP, TCP, UDP) use **check bits in IP header** to detect errors (or in the header and data for TCP/UDP)
- A **checksum is calculated** for header contents and included in a special field.
- Checksum **recalculated at every router**, so algorithm selected for ease of implementation in software
- Let header consist of L , 16-bit words, $\mathbf{b}_0, \mathbf{b}_1, \mathbf{b}_2, \dots, \mathbf{b}_{L-1}$
- The algorithm appends a 16-bit checksum \mathbf{b}_L

Checksum Calculation

The checksum b_L is **calculated as follows**:

- Treating each 16-bit word as an integer, find

$$\mathbf{x} = (b_0 + b_1 + b_2 + \dots + b_{L-1}) \text{ modulo } (2^{16}-1)$$
- The checksum is then given by:

$$\mathbf{b}_L = -\mathbf{x} \text{ modulo } (2^{16}-1)$$

Thus, the headers must **satisfy the following pattern**:

- $$\mathbf{0} = (b_0 + b_1 + b_2 + \dots + b_{L-1} + b_L) \text{ modulo } (2^{16}-1)$$
- The checksum calculation is carried out in software using one's complement arithmetic

Internet Checksum Example

4	0100	• In the receiver	
5	0101	4	0100
9	1001	5	0101
18	10010	9	1001
		-3	1100
18 mod(2 ⁴ -1)	0010	15	11110
= 3	+ 1	15 mod(2 ⁴ -1)	1110
	0011	= 0	+ 1
• Make checksum: -3			1111
• 1100			

Polynomial Codes

- Polynomials instead of vectors for codewords
- Polynomial arithmetic instead of checksums
- Implemented using shift-register circuits
- Also called **cyclic redundancy check (CRC)** codes
- **Most data communications standards use polynomial codes for error detection**
- Polynomial codes also basis for powerful **error-correction** methods

General Idea

- Choose a special code: **G** (generator code, $n=m+k$ bits)
 - Shift information by m bits, $\div G$, and find remainder, **R**
- $$\frac{2^m I}{G} = Q \oplus \frac{R}{G}$$
- Make $n=m+k$ bit codeword
- $$B = 2^m I \oplus R$$
- ↗
m-bit redundancy
- At receiver if **no error**:
- $$\begin{aligned} \frac{B}{G} &= \frac{2^m I \oplus R}{G} \\ &= Q \oplus \frac{R}{G} \oplus \frac{R}{G} = Q \end{aligned}$$
- At receiver if **have error**:
- $$\begin{aligned} \frac{B \oplus E}{G} &= \frac{2^m I \oplus R \oplus E}{G} \\ &= C \oplus \frac{S}{G} \neq Q \end{aligned}$$

Cyclic Error Correction

- We can do more than just detect...

- If have error:

$$\frac{B \oplus E}{G} = \frac{2^m I \oplus R \oplus E}{G} = C \oplus \frac{S}{G}$$

- But note:

$$\frac{B \oplus E}{G} = Q \oplus \frac{E}{G} = C \oplus \frac{S}{G}$$

- *Rearranging:*

$$\frac{E}{G} = [Q \oplus C] \oplus \frac{S}{G}$$

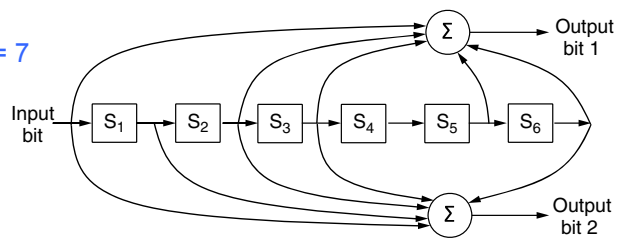
- remainder (**syndrome**) depends only on the error (not on codeword **B**)
- Syndrome can be used to identify error
- As simple as LUT

Cyclic Code Types

- Cyclic codes are a type of **block code**
 - redundant bits are generated by some block of data (contrast with convolutional code)
- BCH codes are a specific example
 - (n,k,d)
 - (7,4,3): code rate = 4/7 = 0.571 (2 detect, 1 correct)
 - (15,5,7): code rate = 5/15 = 0.333 (6 detect, 3 correct)
- Reed-Solomon
 - operate on k-bit symbols (rather than individual bits)
 - and 2^k-1 symbols at a time (e.g. 8-bit symbol & 255 symbols total)
 - typical: (255,233,33), therefore can correct (33 - 1)/2 = 16 symbols
 - 8 x 16 = 128 bits in a 8 x 255 = 2040 bit sequence
 - very good for **burst errors** (DSL, cable, satellite, CDs)

Convolutional Codes

- Codes continuously
 - good for **streaming**, don't have to pause to collect blocks of bits
- Data is shifted through registers
 - output depends on present and past inputs (state-machine)
 - this redundancy achieves the necessary coding
- NASA convolutional code (Voyager)
 - $(2,1)$, $r = \frac{1}{2}$
 - constraint length = 7
 - GSM, 802.11
- Trellis decoding
 - Viterbi algorithm



[Tanenbaum11]

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Recent Iterative Codes

- Turbo codes, 1993
 - two codes generated and interleaved
 - two decoders work iteratively to decode message
 - close to Shannon limit
- Low Density Parity Check, 1962 & 2003
 - block code
 - each output bit formed from only a fraction of input bits
 - iteratively re-assembled
 - rapidly being incorporated (no IP issues)
 - digital video, 10 Gbps ethernet, power line, latest 802.11

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